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TITLE THE NHMFL 60 TESLA, 100 MILLISECOND MAGNET

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THE NHMFL 60 TESLA, 100 MS PULSED MAGNET

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1. INTRODUCTION

Among the new facilities to be offered by the National Science Foundation through the National High Magnetic Field Laboratory (NHMFL) are pulsed fields that can only be achieved at a national user facility by virtue of their strength, duration, and volume. In particular, a 44 mm bore pulsed magnet giving a 60 tesla field for 100 ms is in the final design stage. This magnet will be powered by a 1.4 GW motor-generator at Los Alamos and is an important step toward proving design principles that will be needed for the higher field quasi-stationary pulsed magnets that this power source is capable of driving.

The requirements for the magnet are

- produce a 60T flat-top pulse for 100 ms;
- have a recycle time of approximately 1 hour;
- provide a 77K bore of 44 mm;
- achieve field homogeneity of 10^{-3} or better in 10 mm sphere;
- be robust and reliable, with a lifetime of 10 years or 10,000 pulses for the outer coils and a lifetime exceeding 1000 pulses for the inner coils;
- permit higher field upgrades as better conductors or reinforcements become available;
- be operational by early 1995.

These requirements lead to a poly-coil design consisting of several mechanically independent and spatially separated coils with external reinforcing shells. Among the advantages:

- The reinforcement shell can be customized for each coil as needed to contain stress. This applies to both the hoop and axial stresses.
- Different conductors can be used in different regions.
- Independent power supplies can achieve fast rise times at lower voltage.
- Coils can be individually sized for efficiency without sacrificing homogeneity.
- Faster cooling occurs with separated coils.

- Individual coils are easily replaced to repair damage or install upgrades. (A possible upgrade path using a stronger conductor would be to substitute conductor turns for reinforcement thickness in one or more inner coils.)
- Conductive reinforcing shells can absorb energy from fast field transients caused by faults.
- Failure may be confined to a small number of coils.

A disadvantage to the poly-coil approach is the loss of packing fraction and the consequent need for greater power. However, the NHMFL motor-generator removes available power as a design constraint and cost attention is devoted more to power control and conversion.

To permit timely delivery of the magnet, it was decided to design with conductor and reinforcement materials now commercially available. Tests by NHMFL have confirmed that GlidCop-60 and GlidCop-15, dispersion strengthened copper alloys manufactured by SCM Metal Products, Inc., are adequate conductors for the 60T magnet.

2. MECHANICAL DESIGN AND STRESSES

The design evolved by manually iterating test solutions based on both discrete and averaged mechanical properties of the conductor-insulator coils in contact with reinforcing shells. A more automated iteration method was not attempted because of the difficulty of realistically including discrete and often unquantifiable design considerations related to fabrication, voltage-current, and thermal constraints.

For the most part, the design method follows that used by Lontai and Marston [1] to design the McGill University 10T magnet. This essentially treats the coil and shell as separate continua in a linear background field distribution at the midplane and solves the elastic equations for the hoop stress in the presence of a uniformly distributed Lorentz force in the coil. Only the midplane is treated. Other programs calculate the axial forces which can be included to yield the effective von Mises stress at the midplane. The effect of axial forces was included using results of Markiewicz *et al* [2] which show how axial stress is efficiently transferred to the reinforcing shell, especially in thin coils.

Drawings of the magnet in perspective and cross-section are shown in Figs. 1 and 2. The total conductor mass is 5293 kg and the total reinforcing shell mass is 1527 kg.

The mounting assembly makes provision for aligning the magnetic centers of the 8 coils. The final alignment occurs automatically as the coils are energized and seek their position of lowest energy by moving against leaf springs in the mount. This assembly is shown in Fig. 3 and the assembly in the LN dewar is shown in Fig. 4.

Table 1 contains the mechanical specifications of the 8 coils. Note that the estimated packing fraction realistically takes account of the inevitable winding error.

In studying the stress it was found that thermal loading of some of the windings contributed to stress containment. That is, pre-stressing the reinforcement shell by thermal expansion of the coil (or thermal contraction of the shell during cool-down) effectively loads the coil against the shell and results in smaller coil magnetic stress during the pulse. The problem is illustrated in Fig. 5 which shows the midplane hoop elastic stress without thermal pre-stressing. Note that the stress of the innermost layer of coil 2 (denoted by "E") lies at or above the elastic limit (denoted by the horizontal line labeled E on the right). At the same time the stress in the shell of coil 2 is far below its yield limit (denoted by the horizontal line labeled S on the right). The effect of thermal pre-stressing is to raise the stress in the shell and lower it in the coil. The requirement for robustness has driven the stress design to stay below yield for maximum field pulses, although during the training period some excursions to yield may occur, again for the purpose of obtaining pre-stress.

The midplane radial displacements occurring in the absence of pre-stressing are shown in Fig 6 as cm and in Fig. 7 as percentages. These displacements overestimate the displacement with pre-stress. Even so, the elongation is less than 0.5% for most coils.

The total axial stress on the midplane of each layer are shown in Fig. 8. This is the Lorentz axial stress that takes no account of sharing the stress with the shell or other layers through shear. Likewise, Fig. 9 gives the axial forces per turn for the inner coils where it is strongest.

When account is taken of the thermal pre-stressing and the transfer of axial stress (in the manner described by Markiewicz *et al* [2]) the resulting stresses lie within yield, as listed in Table 2 and shown in Figs. 10 and 11 for interface pressure and thermal stress and the winding and shell stresses, respectively.

3. MATERIALS

Realization of high field non-destructive magnets is material~ limited: conductors of high strength, high conductivity, high specific heat, and good elongation; reinforcement material of high strength at 77K and superior fatigue strength; insulation of low compressibility, high viscosity and high thermal conductivity. The commercial materials chosen for the magnet are as follows.

Conductor: (Al-15 is used for coil 1, AL-60 for coils 2 through 8.)

	<u>GlidCop AL-15</u>	<u>GlidCop AL-60</u>
yield strength (0.2%) @ 77 K	594 MPa	745 MPa
ultimate strength @ 77 K	684 MPa	856 MPa
Young's modulus	139 GPa	143 GPa
elongation @ RT	12.5 %	9.2 %
elect. resistance ratio (77 K/RT)	4.84	3.98
%IACS conductivity @ RT	88.7	79.5
total thermal contraction, RT to 77 K	0.278 %	0.190 %

Reinforcement:

	<u>Nitronic 40 (annealed)</u>
yield strength (0.2%) @ 77 K	1.0 GPa
ultimate strength @ 77 K	1.4 GPa
fatigue strength (1000 cycles) @ 77 K	1. GPa
Young's modulus @ 77 K	186 GPa
elongation @ 77 K	23 %
elect. resistivity @ 77 K	63 micro-ohm-cm
thermal conductivity @ 77 K	7 W/mK
total thermal contraction, RT to 77 K	0.259 %

Insulation:

	<u>CTD-101G</u>	<u>CTD-101 w/ 50% g</u>
compression strength @ 77 K	560 MPa	1.25 GPa
compression modulus @ 77 K	17.7 GPa	
shear strength @ 77 K	170 MPa	200 MPa
shear modulus @ 77 K	8.2 GPa	9.1 GPa
thermal conductivity @ 77 K	10-20 W/mK	
total thermal contraction, RT to 77 K	0.4%	
viscosity @ 110 °C	2,000 cP	

The CTD-101G resin is 67% alumina by weight and has a remarkably high thermal conductivity. This makes it possible to cool coil 8 in one hour. The characteristics for CTD-101, which can be used for the thin coils, 1 through 7, refer to the impregnated composite of resin and 50% S-2 fiberglass. The conductor will be half-lapped with fiberglass and Kapton tape before being wound on the mandrel. The coils will then be potted in the usual manner, with the reinforcing shell used as part of the impregnation mold for some of the coils.

4. ELECTRICAL BEHAVIOR

4.1 Results and Parameters

The lengths of the inner windings were chosen to reach 95% field efficiency (compared to an infinitely long coil) and to remain within homogeneity limits. This is illustrated in Fig. 12. The 1-cm homogeneity is predicted to be $1.3 \cdot 10^{-4}$. Contours of the field strength are shown in Fig. 13. These fields are used to predict the force on current leads to the coils. The midplane magnetic profile is given in Fig. 14.

Table 3 lists the electrical parameters that give the above fields. Note that the identical currents in coils 1 through 7 do not imply a common power supply.

4.2 Electrical Circuit

The design of the power supply for the 60 T magnet is a compromise between many parameters, such as coil stress and heating, modularity, upgrade capability and cost. The inner coils are designed for a shorter current pulse than the outer coil. Therefore, the coils are partitioned electrically into several groups. Although the seven inner coils have the same peak current, they must be partitioned further into two groups to avoid overheating the inner coils: the inner group comprising coils 1, 2, and 3 and the intermediate group of coils 4, 5, 6 and 7. Coil 8 is treated independently. Because of the strong influence of the mutual coupling, it was decided to keep the current in the outer groups nearly constant while the current in the inner group ramps up (Fig. 15). If the current in the inner coil group is ramped up while the outer groups are still ramping, a considerably higher voltage must be applied to maintain the ramp rate in the outer coils.

Three power supplies are used for the three groups: coils 1-3, 4-7, and 8, respectively. Table 4 gives the peak current, I_p , voltage at the peak current, V , and the peak power requirement, P . The values of Table 4 are taken from simulation results. Coil 3 was included in the inner coil group to have enough inductance in the circuit to obtain an acceptable ripple value during the 60T, 100 ms flat top.

Given the peak current values and the required voltage values at peak current, the no-load voltage of the each supply can be determined. Assuming a 16% voltage drop between no-load and full-load voltage, the following no-load voltage, V_o , is obtained:

Supply 1	$V_o \approx 1.2 \text{ kV}$
Supply 2	$V_o \approx 8.0 \text{ kV}$
Supply 3	$V_o \approx 12.0 \text{ kV}$

A 16% voltage drop for pulsed converters is a reasonable assumption. Supplies with such a voltage drop have been built for pulsed fusion experiments. A power supply with a no-load voltage in the range of 10 kV must be built up from several modules, which have no-load voltages in the range of 3 to 4 kV. Let us propose the higher voltage 4 kV design. The outer coil supply will need three modules, connected in series. The intermediate coils will need two modules connected in series. Each module is a 12 pulse unit, consisting of two series-connected 6 pulse units. Two 12 pulse modules can be arranged as one 24 pulse unit. However, because of the large inductance, the additional circuit complexity does not justify the small improvement in the ripple. Fig. 16 (in the text) shows the supply module arrangement for the outer coil and the intermediate coils.

A module consists of a 24 kV circuit breaker, a three winding transformer, two series connected 6 pulse bridges and two 'thyristorised' crowbar paths. Each bridge has an output voltage of 2 kV and an input voltage of 1.5 kV. While the voltage rating of the two major supplies can be satisfied with identical modules, the current rating is different, because the current pulse length is different. The supply for the outer coil must be designed for a trapezoidal current with a 1.6 s rise and fall time and a 0.8 s flat top-time. Assuming a flat-top current

Table 4. Power Supply Parameters

	I_p	V	P
Supply 1 (coil 1, 2, 3)	18.64 kA	10.0 kV	18.64 MW
Supply 2 (coil 4, 5, 6, 7)	18.64 kA	6.6 kV	123.00 MW
Supply 3 (coil 8)	16.52 kA	10.0 kV	165.00 MW

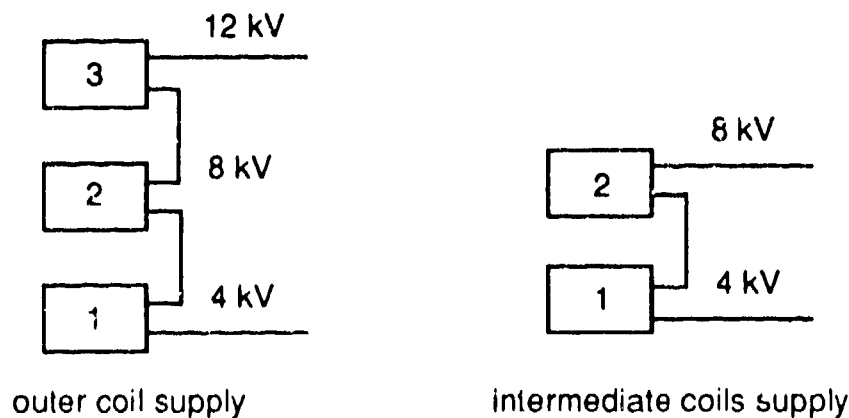


Figure 16. Power supply module arrangement for coil groups.

of 17 kA, the supply current must be dimensioned for an I^2t rating of $540 \cdot 10^6$

A²S. An equivalent rectangular current with an amplitude of 17 kA and a pulse length of 2 s (1.866 s, more precisely) has the same heating effect.

The supply for the intermediate coils has a rise and decay time of about 0.3 s, a flat-top time of 0.225 s and a flat-top current of 18.6 kA. The equivalent rectangular pulse current with a 17 kA amplitude has a pulse length of 0.4s. If minimum cost is the only determining factor for purchasing the power supplies, NHMFL will have two different types of power supply modules for the 60T magnet. Considering the fact that a facility is being built which should supply power for magnets above 60T with increased power requirements the preferred choice is to buy 12 pulse, 4 kV, 20 kA, 2 s pulse length modules. These modules can be connected in parallel and series to accommodate different requirements.

The inner coil group will be supplied by an existing 12 pulse power supply. Each power supply should be equipped with a free wheeling path, made of thyristors.

4.3 Electrical Ripple

Power supplies generate voltage ripple when the bridge converters are phased back, resulting in a lower average output voltage. At the beginning of the flat top the average coil voltages are the smallest, resulting in the highest ripple voltages. If the simplest converter control strategy is used to achieve the reduction in average voltage, the ripple current can be calculated to be $2.5 \cdot 10^{-3}$, $0.3 \cdot 10^{-3}$ and $5.5 \cdot 10^{-3}$ for the field produced by the inner, intermediate and outer coil group respectively. Table 5 gives the assumption for the ripple calculation with V_o the supply no-load voltage, V the average voltage at the beginning of the flat-top, V_{pp12} and I_{pp12} the peak to peak 12th harmonic voltage and current, L the inductance of the coil group and R the ratio of I_{pp12} to the flat-top current, I_{max} . All the values are approximated. As expected, the inner coil group doesn't have enough inductance to smooth out the current. Some advanced converter control for the inner coil group supply can reduce the ripple to $5 \cdot 10^{-4}$.

Table 5. Ripple Calculation

	V_o	V	V_{pp12}	L	I_{pp12}	I_{max}	R
Supply 1	1.2 kV	0.45 kV	0.7 kV	5.00 mH	41 A	18.64 kA	$2.5 \cdot 10^{-3}$
Supply 2	8.0 kV	4.70 kV	3.0 kV	0.15 H	6 A	18.64 kA	$3.0 \cdot 10^{-4}$
Supply 3	12.0 kV	3.80 kV	8.0 kV	0.80 H	3 A	16.50 kA	$1.7 \cdot 10^{-4}$

5. THERMAL BEHAVIOR

There are two aspects to the thermal behavior, one critical and the other convenient.

The critical thermal behavior is the final coil temperature, because the epoxy insulation degrades at temperatures modestly above room temperature. In this design, no coil reaches such temperatures. Another issue is the heating of the metallic shells due to induced currents during field changes. The predicted temperature rise and the energy deposited in the steel shells are shown in Fig. 17. The winding and shell temperatures after a pulse are summarized in Fig. 18.

The convenient side of thermal behavior is the cool-down time of the magnet, because this fixes the maximum number of shots per day. Coil 8, being the thickest, determines the cool-down. Thanks to the high thermal conductivity of CTD-101G alumina-loaded resin the goal of one-hour cooling can be reached, as shown in Figs. 19, 20, and 21 for different starting temperatures, different thermal conductivities of the insulating resin, and different coil thicknesses, respectively.

6. MOTOR-GENERATOR

6.1 Background

Los Alamos National Laboratory's "1430 MVA Generator Facility" is based on a motor-generator set which in its current configuration is capable of producing high power pulses of up to 1040 MVA of peak power and up to 600 MJ of energy. This same set and the facility could, given the need and the funds, be upgraded to provide pulses of up to about 2000 MVA and 2000 MJ.

Besides the regular, high power, long pulse applications of the motor-generator there are other possible uses, such as:

- 3 - Phase loads. High current, low power, long and very long pulses. (As installed, up to 25 kA; maximum, up to about 50 kA)
- 2 - Phase loads. Very short pulses, as long as one cycle. (As installed, up to 840 MVA apparent power; maximum, up to about 1680 MVA)
- "Short circuit loads" with hardly any voltage, current controlled. (As installed, up to 25 kA; maximum, up to 50 kA)

All of the above applications would require modifications to the existing facility configuration.

6.2 Capabilities of the Generator Facility

The list below gives some of the basic rated data of the motor-generator.

Rated power	1430	MVA
Rated voltage	24	kV
Rated current	34.4	kA
Rated field voltage	670	V
Rated field current	8100	A
Rated frequency	60	Hz
Rated speed	1800	rpm

The following table summarizes some of the capabilities and limitations of the facility for the long pulse regime. This data is valid for the facility as it is configured to date.

Pulse power range	0 - 1040	MVA
Pulse voltage range	0 - 24	kV
Pulse current range	0 - 25	kA
Pulse speed range	1260 - 1800	rpm
Stored energy @ 1800 rpm	1260	MJ
Extractable energy (1800 rpm → 1260 rpm)	600	MJ
Full pulse repetition rate	10	/min
Maximum pulse duration @ 24 kV	10	s
Maximum pulse duration @ 25 kA	2	s
Maximum effective current averaged over a 10 minute period	1.8	kA
Maximum field voltage	1025	V
Maximum field current	5130	A
Run up time (9 rpm → 1800 rpm)	15	min
Run up time between full pulses (1260 rpm → 1800 rpm)	7	min
Run down time (1800 rpm → 9 rpm)	6	min
Coast down time (1800 rpm → 9 rpm)	38	min
Fatigue life: Start / Stop (9 rpm → 1800 rpm → 9 rpm)	1700	Cycles
Full pulses (1800 rpm → 1260 rpm → 1800 rpm)	10 ⁵	Cycles

The above tables show that the facility can easily cover the needs of the NHMFL projects to be realized in the near future (i.e. 60 T magnet system).

6.3 Possible Upgrades

The facility can be upgraded in several different ways, depending on the needs and funds available.

Increase of the stored and extractable energy

With a flywheel the stored energy at 1800 rpm could be increased from 1260 MJ to a maximum of about 4000 MJ. This would increase the extractable energy from 600 MJ to about 2000 MJ. Together with an increase of the pulse power output this will most likely satisfy all possible needs of the magnet facility even for the distant future.

Sharp and at the same time high magnitude load transitions will significantly shorten the fatigue life of the shaft train in this configuration (Power transients of about 1500 MW in 20 ms reduce the fatigue life to about 800 full pulse cycles).

The run up times upon start up and re-acceleration between pulses would be about three to four times longer. The same is true for the run down and coast down times.

The machine set and the foundation are already designed to accommodate a future flywheel. Space for the necessary auxiliary equipment is also reserved. The control system is designed to accommodate the additional control and protection tasks.

Increase of the power output

The pulse power output is at the time limited by the following elements:

- current limiting and interrupting device;
- load breaker;
- excitation transformer.

The current limiting and interrupting device, currently under procurement, will be one of highest rated devices of its kind worldwide (24 kV, 25 kA, 2 s, with a trigger level of 35 to 40 kA momentary current). The use of this element greatly reduces the costs for the 24 kV bus by making it possible to use standard industrial bus designs, high voltage breakers and related equipment. It also limits drastically the extent of damages in case of faults on the primary side of the power converter transformers.

Our experiences with this device, future developments of this product and more elaborate schemes will certainly make even higher ratings possible.

7. SCHEDULE AND COST ESTIMATE

Table 6 lists estimates of sub-task completion dates and times and some costs.

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Tables

- Table 1. Mechanical Specifications.
- Table 2. Stress Results.
- Table 3. Electrical Parameters.
- Table 4. (In text.)
- Table 5. (In text.)
- Table 6. Schedule and Cost Estimates.

Figures

- 1. 60T magnet in perspective.
- 2. 60T magnet in cross-section.
- 3. 60T magnet in mounting assembly.
- 4. 60T magnet and mounting assembly in dewar.
- 5. Hoop stress in the absence of thermal pre-stress.
- 6. Radial displacement on midplane.
- 7. Percentage radial displacement on midplane.
- 8. Total axial stress on midplane layers.
- 9. Axial forces on individual turns.
- 10. Interface pressure and thermal stress.
- 11. Winding and shell stresses.
- 12. Coil lengths compared to saturation and homogeneity lengths.
- 13. Contours of field strength.
- 14. Midplane magnetic profile.
- 15. Desired current pulse in the three coil group.
- 16. (In text.)
- 17. Temperature rise and the energy deposited in the steel reinforcing shells.
- 18. Winding and shell temperatures.
- 19. Cooling for different starting temperatures.
- 20. Cooling for different thermal conductivities of the insulating resin.
- 21. Cooling for different coil thicknesses.

MECHANICAL INFORMATION

Winding	one	two	three	four	five	six	seven	eight
Alpha (Winding OD/Winding ID)	1.53	1.64	1.33	1.20	1.07	1.06	1.05	1.68
Beta (Winding Height/Winding ID)	4.26	4.19	3.55	3.33	2.28	1.92	1.65	1.45
Winding ID (cm)	4.75	9.06	18.3	30	43.8	52.16	60.52	69.08
Winding OD (cm)	7.26	14.9	24.4	36.1	46.86	55.22	63.58	116.32
Winding Radial Thickness (cm)	1.26	2.92	3.05	3.05	1.53	1.53	1.53	23.62
Winding Average Radius (cm)	3.00	5.99	10.68	16.53	22.67	26.85	31.03	46.35
Winding Height (cm)	20.25	38	65	100	100	100	100	100
Reinforcing Shell OD (cm)	8.46	17.7	29.4	43	51.36	59.72	68.28	----
Reinforcing Shell Thickness (cm)	0.6	1.4	2.5	3.45	2.25	2.25	2.35	----
Reinforcing Shell Height (cm)	25.3	41.0	70.1	106.0	106.0	106.0	106.0	----
Reinforcing Shell Mass (kg)	2.94	23.01	115.93	355.78	288.12	337.16	403.99	----
Radial Winding Error per Layer (cm)	0.07	0.07	0.07	0.07	0.07	0.07	0.07	0.09
Axial Winding Error per Turn (cm)	0.03	0.03	0.03	0.04	0.04	0.04	0.04	0.12
Winding Packing Factor	0.65	0.70	0.70	0.70	0.70	0.70	0.70	0.74
Conductor Size Hght x Wdth (cm)	.78x.47	.95x.57	1.0x.6	1.0x.6	1.0x.6	1.0x.6	1.0x.6	1.4x1.0
Conductor Cross-Sect. Area (cm^2)	0.3613	0.5362	0.5947	0.5947	0.5947	0.5947	0.5947	1.3947
IR to Innermost Cu Surface (cm)	2.41	4.58	9.20	15.05	21.95	26.13	30.31	34.59
OR to Outermost Cu Surface (cm)	3.58	7.40	12.15	18.00	23.38	27.56	31.74	58.11
Turns per Layer	23	36	59	90	90	90	90	63
Layers	2	4	4	4	2	2	2	20
Turns (total)	46	144	236	360	180	180	180	1260
Conductor Length (m)	8.7	54.2	158	374	256	304	351	3669
Conductor Mass (kg)	2.76	25.60	83	196	134	159	184	4509
Conductor Alloy	AL-15	AL-60	AL-60	AL-60	AL-60	AL-60	AL-60	AL-60
Combined Winding & Shell Mass (kg)	5.70	48.62	199	552	422	496	588	4509
Total Conductor Mass (kg) =	5293		Conductor corner radii are 0.0787 cm,					
Total Reinforcing Shell Mass (kg) =	1527		0.3 cm wide annuli between windings 1 through 4,					
Total Metal Mass (kg) =	6820		0.4 cm wide annuli between windings 4 through 8.					
			Winding 1 lined with a 0.15 cm wall thickness					
			epoxy fiberglass laminate tube to protect winding.					

Table 1.

STRESS INFORMATION

Winding	one	two	three	four	five	six	seven	eight
Bo (T)	60	54.9	46.44	38.363	30.356	26.516	22.801	19.218
delta B (T)	5.1	8.46	8.077	8.007	3.84	3.715	3.583	19.218
Beta	4.263	4.194	3.552	3.333	2.283	1.917	1.652	1.448
Wndng ID (cm)	4.75	9.06	18.3	30	43.8	52.16	60.52	69.08
Wndng OD (cm)	7.26	14.9	24.4	36.1	46.86	55.22	63.58	116.32
delta r (cm)	1.255	2.92	3.05	3.05	1.53	1.53	1.53	23.62
r avg (cm)	3.0025	5.99	10.675	16.525	22.665	26.845	31.025	46.35
Length (cm)	20.25	38	65	100	100	100	100	100
OD Spprt (cm)	8.46	17.7	29.4	43	51.36	59.72	68.28	----
Spprt Thk (cm)	0.6	1.4	2.5	3.45	2.25	2.25	2.35	----
Packing Fctr	0.654	0.696	0.708	0.702	0.7	0.7	0.7	0.744
Cndctr Sz H x W (cm)	.78x.47	.95x.57	1.0x.6	1.0x.6	1.0x.6	1.0x.6	1.0x.6	1.4x1.0
Cndctr X-sect area (cm^2)	0.3613	0.5362	0.5947	0.5947	0.5947	0.5947	0.5947	1.3947
IR to Cu Surf (cm)	2.4106	4.5808	9.2008	15.051	21.951	26.131	30.311	34.591
OR to Cu Surf (cm)	3.5792	7.3992	12.149	17.999	23.379	27.559	31.739	58.109
Turns (total)	46	144	236	360	180	180	180	1260
Layers	2	4	4	4	2	2	2	20
Cndctr Length (m)	8.678	54.196	158.29	373.79	256.33	303.61	350.88	3669.4
j (kA/cm^2)	51.6	34.76	31.34	31.34	31.34	31.34	31.34	11.85
I (kA)	18.64	18.64	18.64	18.64	18.64	18.64	18.64	16.52
Peak I (kA)	18.64	18.64	18.64	18.64	18.64	18.64	18.64	17.2*
Winding Temp I (K)	97	92	87	155	155	155	155	----
Winding Temp F (K)	150	128	113	196	196	196	196	----
Shell Flat Top Temp (K)	77.3	78.1	79.6	81.3	82.5	83.3	83.2	
Raw Wnd Axl Strs (MPa) I	12	43	135	364	429	464	495	198*
Rw Wnd Axl Strs (MPa) Av	12	47	141	373	433	468	498	----
VM Stress Wndng (MPa) I	521	680	687	581	469	487	459	689*
VM Stress Spprt (MPa) I	732	918	907	1050	955	978	945	----
Rad Intrfc Press (MPa) I	123	168	170	173	82	72	63	----
Rad Intrfc Press (MPa) F	142	180	177	180	86	75	65	----
VM Stress Wndng (MPa) F	436	635	652	525	416	434	407	----
VM Stress Spprt (MPa) F	863	988	945	1095	987	1009	975	----
VW Th Strs Wndng (MPa) I	122	162	110	209	204	201	202	----
VW Th Strs Shell (MPa) I	190	262	116	165	116	113	108	----
VW Th Strs Wndng (MPa) F	205	207	146	268	263	259	260	----
VW Th Strs Shell (MPa) F	325	333	154	211	150	146	140	----
I= Flat Top Start, F= Flat Top End				* Overdrive to 20 T				

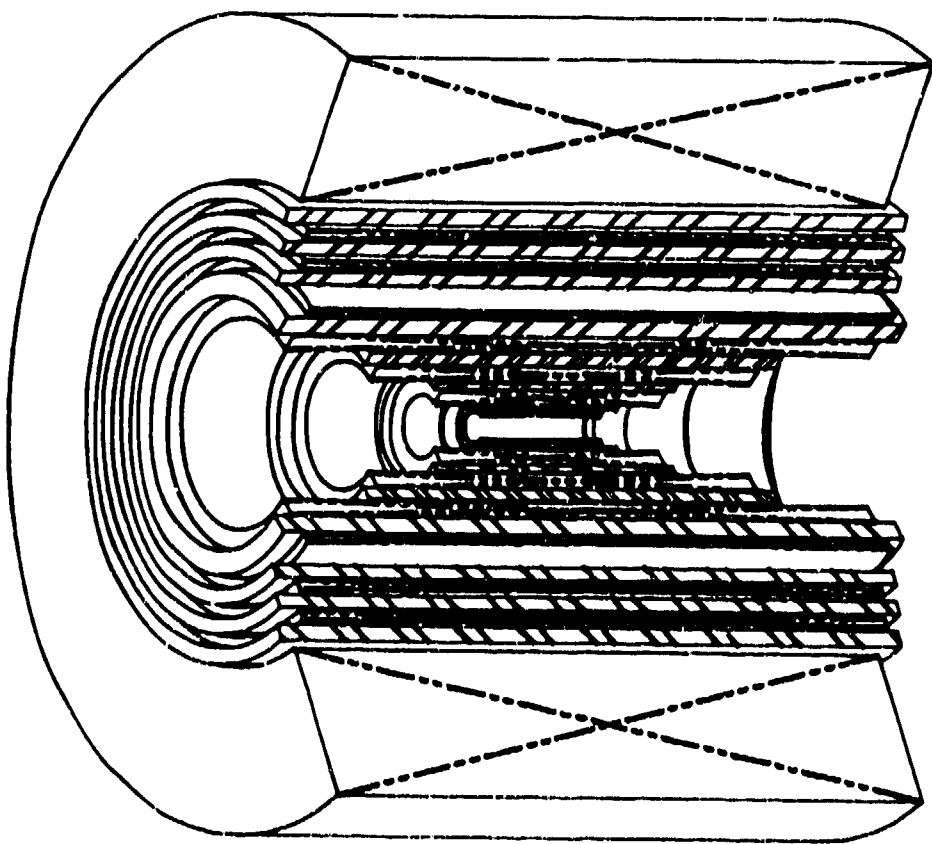
Table 2.

Winding	one	two	three	four	five	six	seven	eight
Winding Effective Center Field (T)	60	54.9	46.44	38.363	30.356	26.516	22.801	19.218
Field Produced by Winding (T)	5.1	8.46	8.077	8.007	3.84	3.715	3.583	19.218
Alpha (Winding OD/Winding ID)	1.53	1.64	1.33	1.20	1.07	1.06	1.05	1.68
Beta (Winding Height/Winding ID)	4.26	4.19	3.55	3.33	2.28	1.92	1.65	1.45
Fraction of Max. Possible Field (%)	95	95	95	95	92	88	84	74
Winding ID (cm)	4.75	9.06	18.3	30	43.8	52.16	60.52	69.08
Winding OD (cm)	7.26	14.9	24.4	36.1	46.86	55.22	63.58	116.32
Winding Height (cm)	20.25	38	65	100	100	100	100	100
Winding Packing Factor	0.65	0.70	0.70	0.70	0.70	0.70	0.70	0.74
Conductor Size Hght x Wdth (cm)	.78x.47	.95x.57	1.0x.6	1.0x.6	1.0x.6	1.0x.6	1.0x.6	1.4x1.0
Conductor Cross-Sect. Area (cm ²)	0.3613	0.5362	0.5947	0.5947	0.5947	0.5947	0.5947	1.3947
Turns per Layer	23	36	59	90	90	90	90	63
Layers	2	4	4	4	2	2	2	20
Turns (total)	46	144	236	360	180	180	180	1260
Self Inductance (mH)	0.0285	0.574	3.04	11.4	5.34	7.28	9.45	763
Mutual Inductance w/ Winding 1 (mH)		0.0558	0.0566	0.0563	0.027	0.0261	0.0251	0.152
Mutual Inductance w/ Winding 2 (mH)			0.704	0.705	0.337	0.325	0.313	1.89
Mutual Inductance w/ Winding 3 (mH)				3.57	1.69	1.63	1.56	9.41
Mutual Inductance w/ Winding 4 (mH)					5.71	5.46	5.24	31.8
Mutual Inductance w/ Winding 5 (mH)						5.21	4.99	30.1
Mutual Inductance w/ Winding 6 (mH)							7.09	42.6
Mutual Inductance w/ Winding 7 (mH)								57.5
Resistance at Pulse Start (Ohms)	0.0007	0.0051	0.0135	0.0315	0.0215	0.024	0.026	0.047
Resistance at Pulse End (Ohms)	0.0028	0.0128	0.0218	0.1085	0.074	0.0875	0.105	0.325
Pulse Length (seconds)	0.494	0.494	0.494	0.837	0.837	0.837	0.837	4.25
Current Density in Cond. (kA/cm ²)	51.6	34.76	31.34	31.34	31.34	31.34	31.34	11.85
Current in a Turn (kA)	18.64	18.64	18.64	18.64	18.64	18.64	18.64	16.52
Peak Current in a Turn (kA)	18.64	18.64	18.64	18.64	18.64	18.64	18.64	17.2
Resistive Energy (MJ)	Windings 1, 2 & 3 = 2.63			Windings 4, 5, 6 & 7 = 33.5				123
Peak Total Energy (MJ)	Windings 1, 2 & 3 = 3.7			Windings 4, 5, 6 & 7 = 70				290
Peak Power (MW)	Windings 1, 2 & 3 = 19			Windings 4, 5, 6 & 7 = 113				168

Table 3.

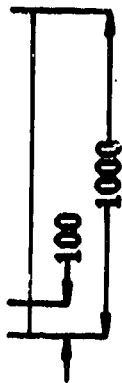
MAGNET SCHEDULE & COST ESTIMATE

	<u>TASK</u>	<u>TIME REQ'D</u>	<u>LATEST START</u>	<u>COST K\$</u>
1.	Material testing	20 weeks	1 FEB 93	
2.	Prototype testing	30 weeks	15 NOV 92	100
3.	Design & analysis	26 weeks	15 OCT 92	
4.	Prepare drawings	8 weeks	1 MAY 93	
5.	Obtain dewar fab. contract	18 weeks	1 DEC 93	
6.	Fab. dewar	20 weeks	1 APR 94	100
7.	Obtain cond. fab. contract	16 weeks	15 JUL 93	
8.	Fab. conductor	18 weeks	1 DEC 93	200
9.	Obtain shell fab. contract	18 weeks	1 JUL 93	
10.	Fab. shells	20 weeks	15 NOV 93	200
11.	Obtain coil fab. contract	18 weeks	1 DEC 93	
12.	Fab. coil	20 weeks	15 APR 94	200
13.	Obtain frame fab. contract	18 weeks	1 MAR 94	
14.	Fab. frame	18 weeks	1 JUN 94	100
15.	Install & test magnet	12 weeks	1 SEP 94	
16.	System commissioning	18 weeks	1 DEC 94	
17.	60 T system ready		1 APR 95	



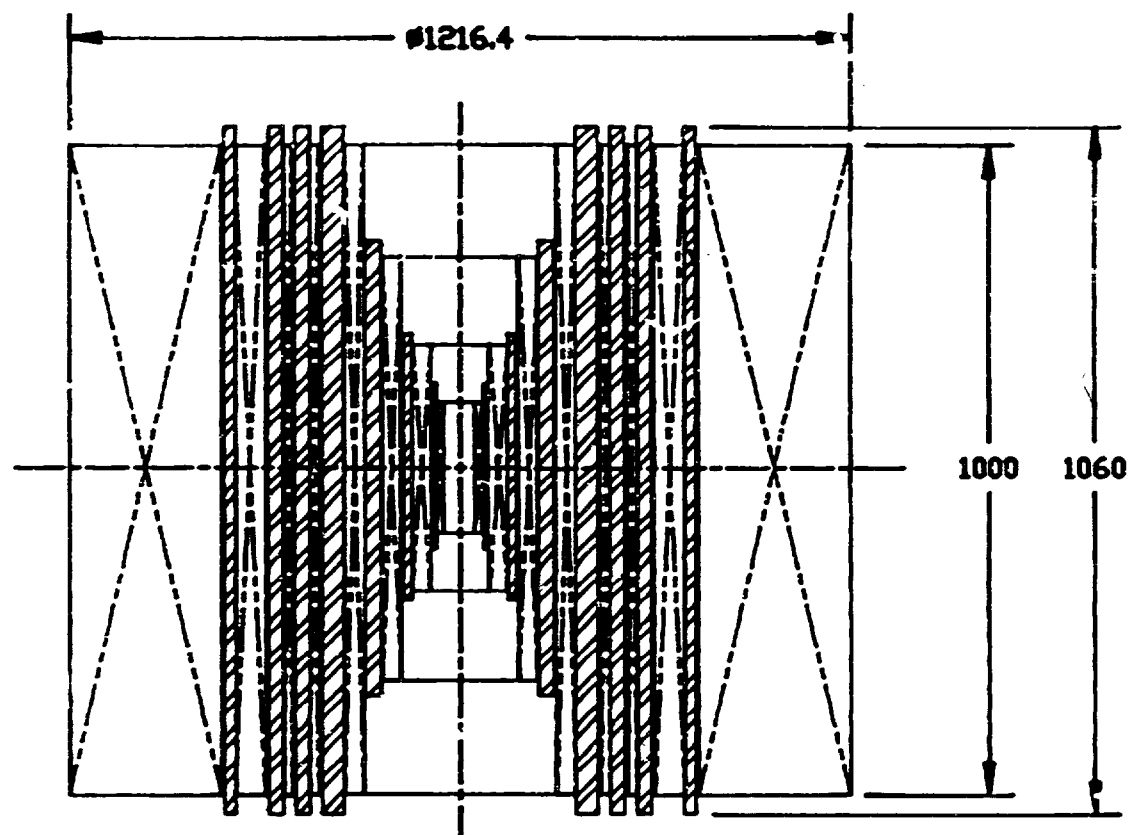
60 TESLA 100ms MAGNET WINDINGS & SHELLS

GROUP M6/LDS ALALMOS NATIONAL LABORATORY
NATIONAL HIGH MAGNETIC FIELD LABORATORY



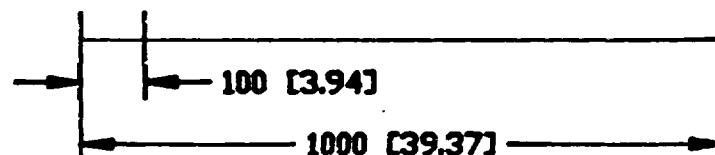
Dimensions are in MILLIMETERS
Dimensions in Brackets are in INCHES

Fig. 1

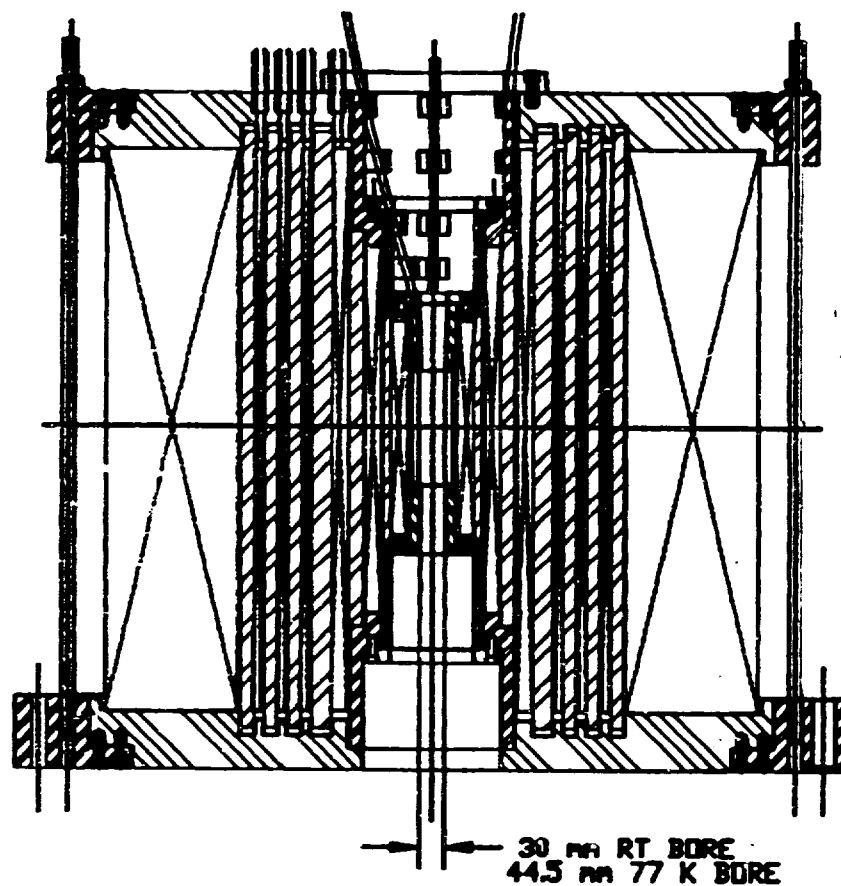


60 TESLA 100 ms MAGNET WINDINGS & SHELLS

GROUP M6/LDS ALAMOS NATIONAL LABORATORY
NATIONAL HIGH MAGNETIC FIELD LABORATORY

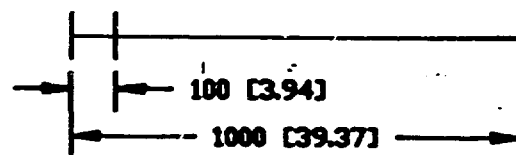


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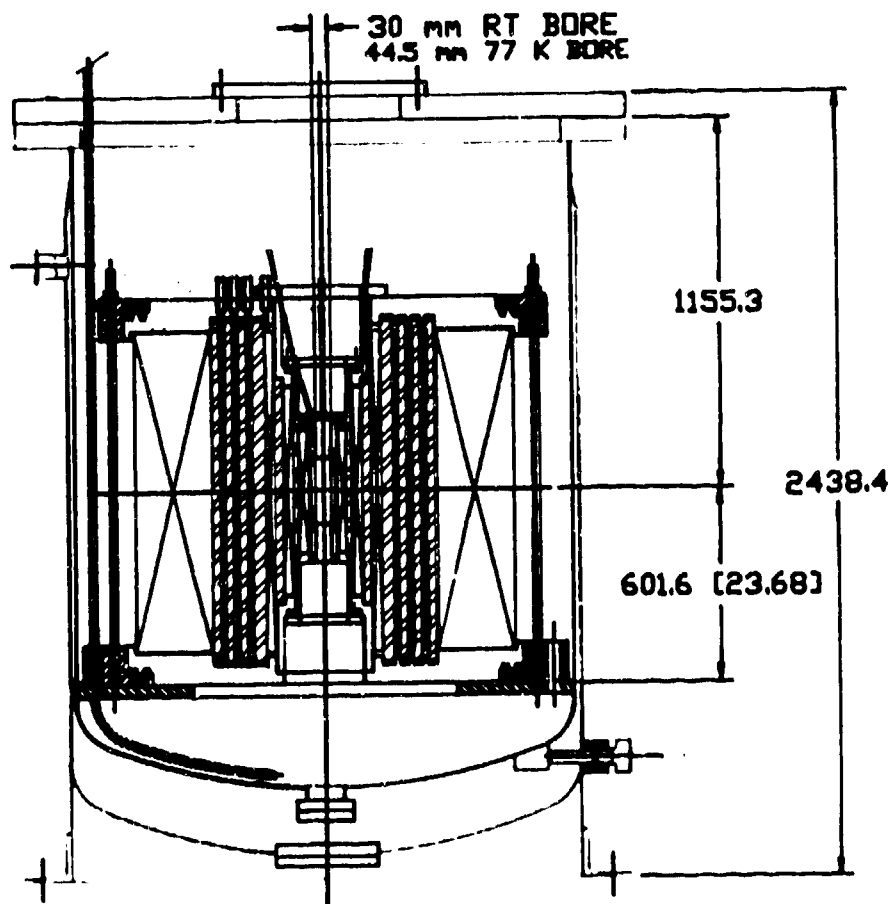


60 TESLA 100ms MAGNET

GROUP M6/LDS ALALMOS NATIONAL LABORATORY
NATIONAL HIGH MAGNETIC FIELD LABORATORY

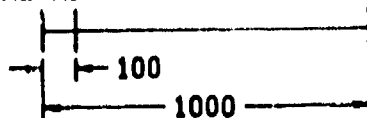


Dimensions are in MILLIMETERS
Dimensions in Brackets are in INCHES



60 TESLA 100ms MAGNET & DEWAR

GROUP M6/LDS ALAMOS NATIONAL LABORATORY
NATIONAL HIGH MAGNETIC FIELD LABORATORY



Dimensions are in MILLIMETERS
Dimensions in Brackets are in INCHES

Fig. 4

Hoop Elastic Stress (w/o thermal loading)

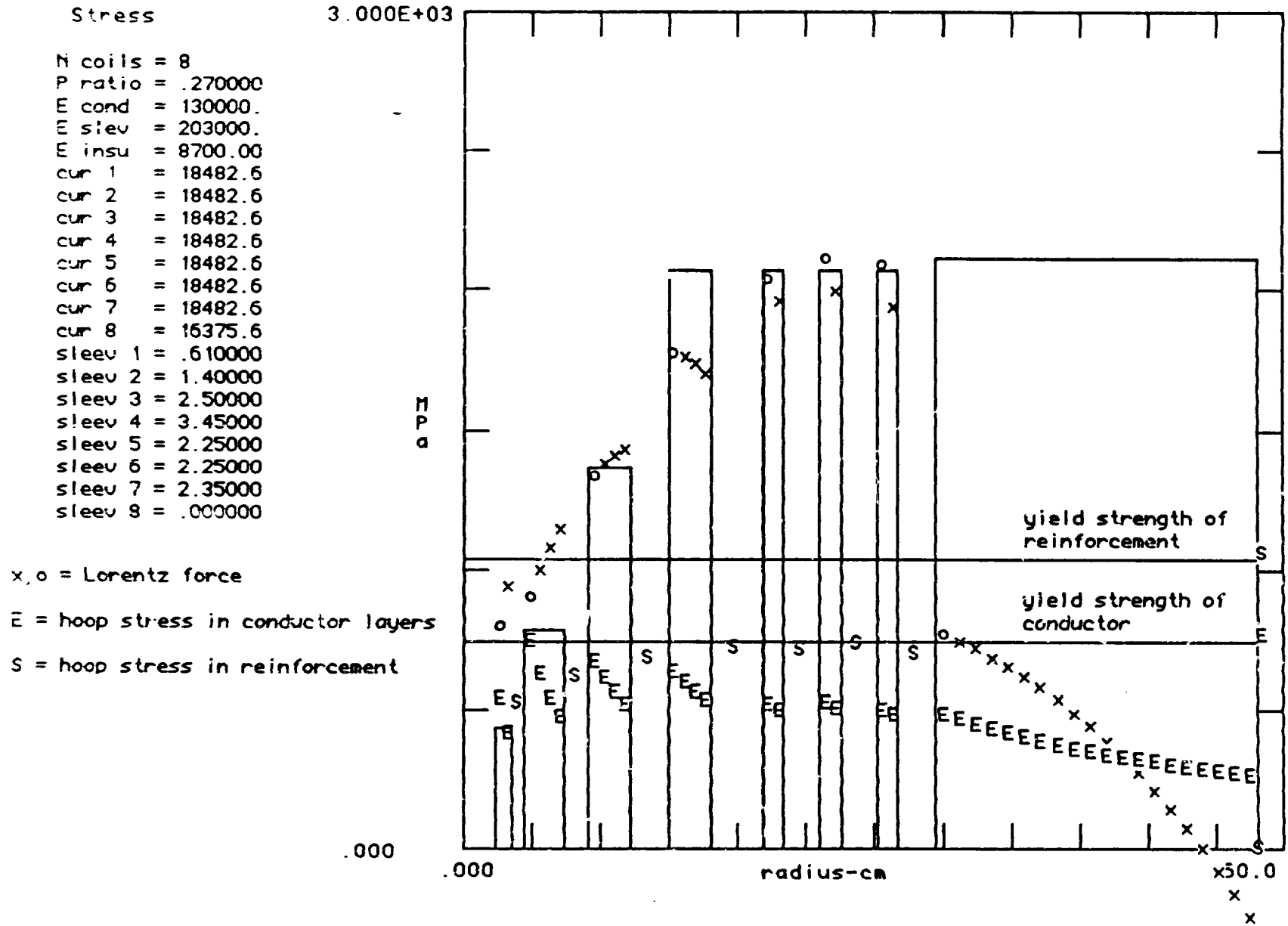
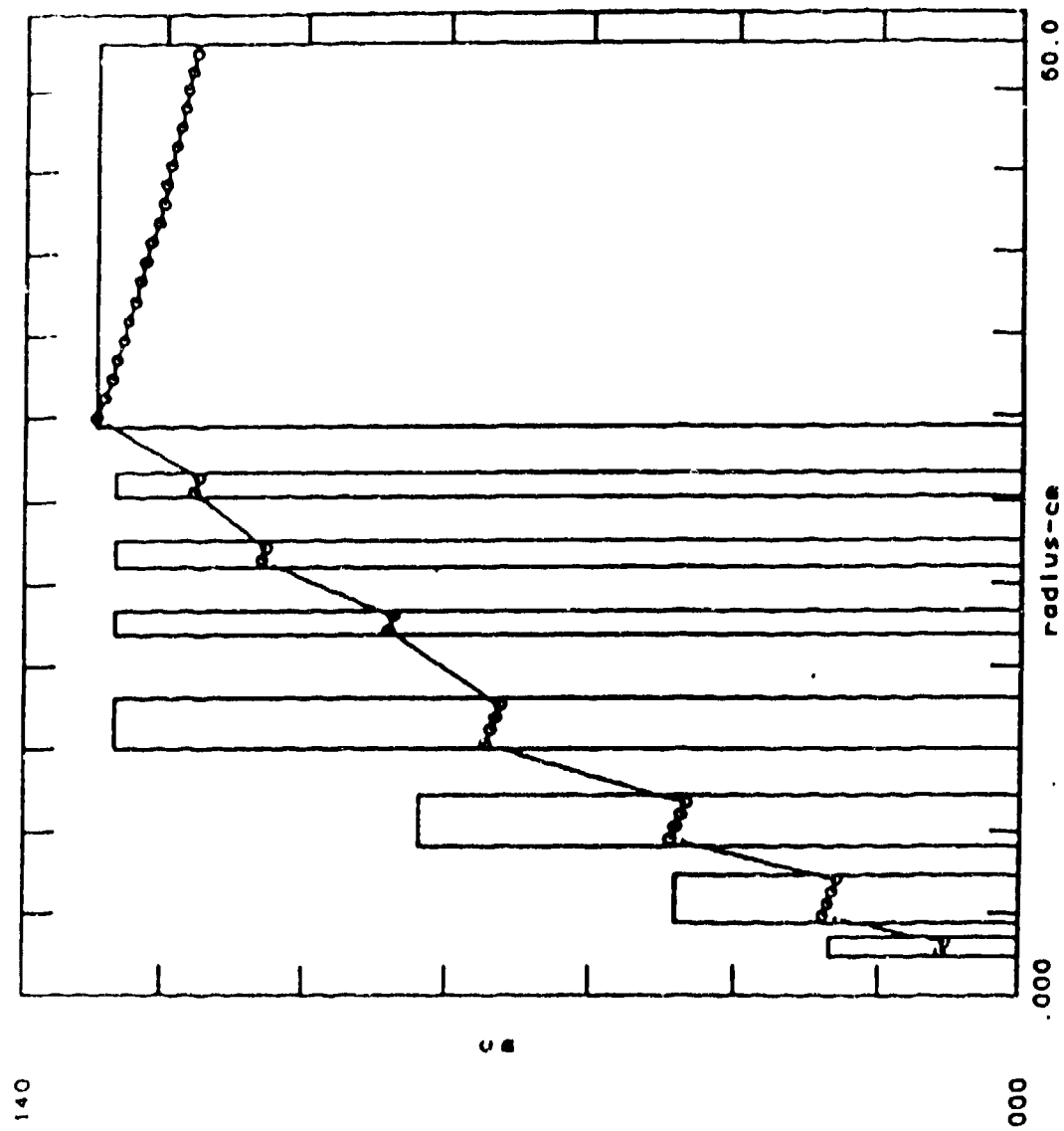


Fig. 5

Radial Displacement from Hoop Stress



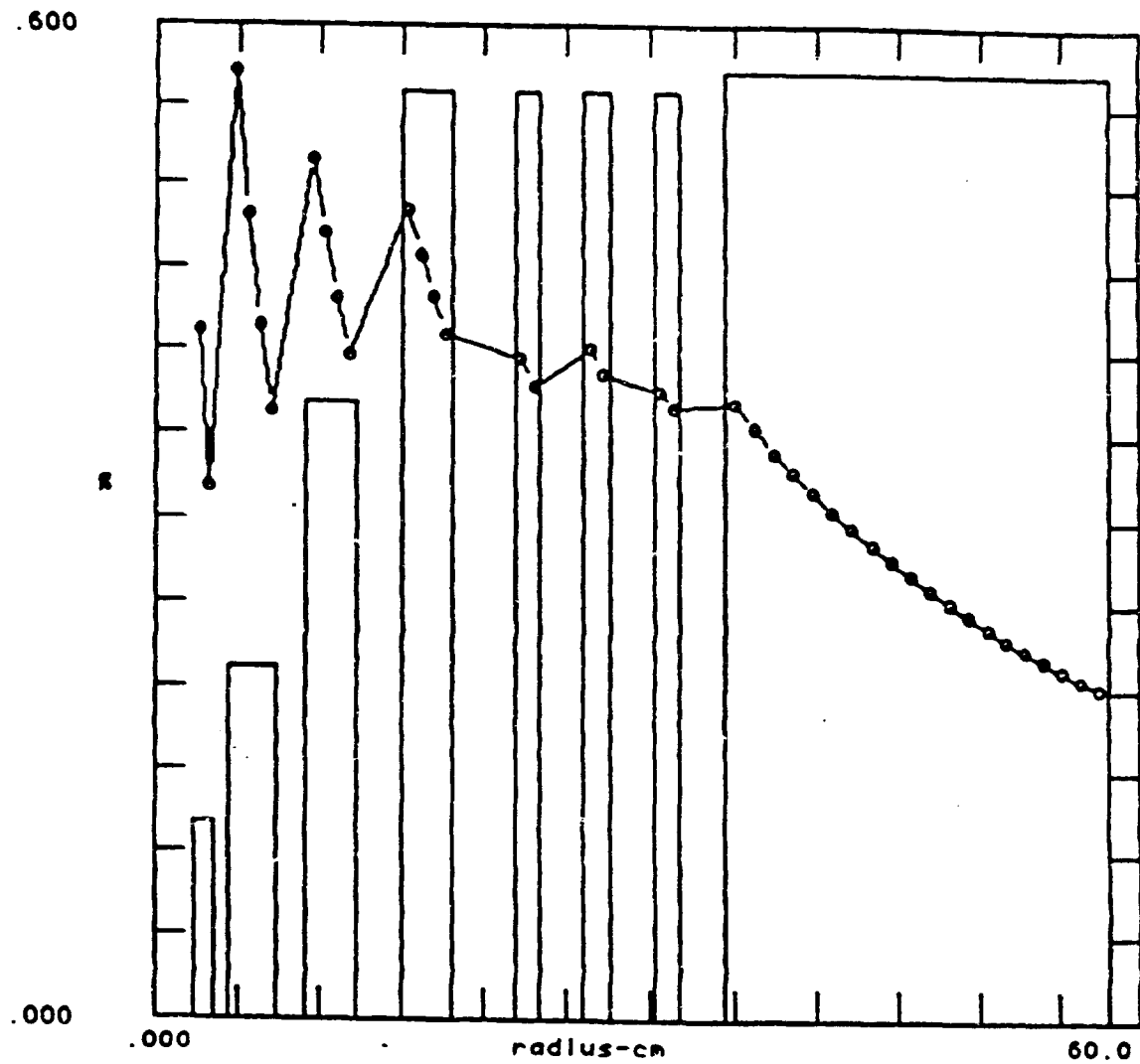
C displacement
 ur(1) = 470000
 uz(1) = 769000
 ur(end) = 1.00000
 uz(end) = 1.39500
 gap(1) =
 .3 3 3 4 .4
 .4 .4 0
 r1 1 = 2.35000
 r1 2 = 4.52000
 r1 3 = 9.14000
 r1 4 = 14.9800
 r1 5 = 21.8700
 r1 6 = 26.0400
 r1 7 = 30.2100
 r1 8 = 34.4800
 r2 1 = 3.61000
 r2 2 = 7.44000
 r2 3 = 12.1800
 r2 4 = 18.0200
 r2 5 = 23.3900
 r2 6 = 27.5600
 r2 7 = 31.7300
 r2 8 = 58.1200

Fig 6

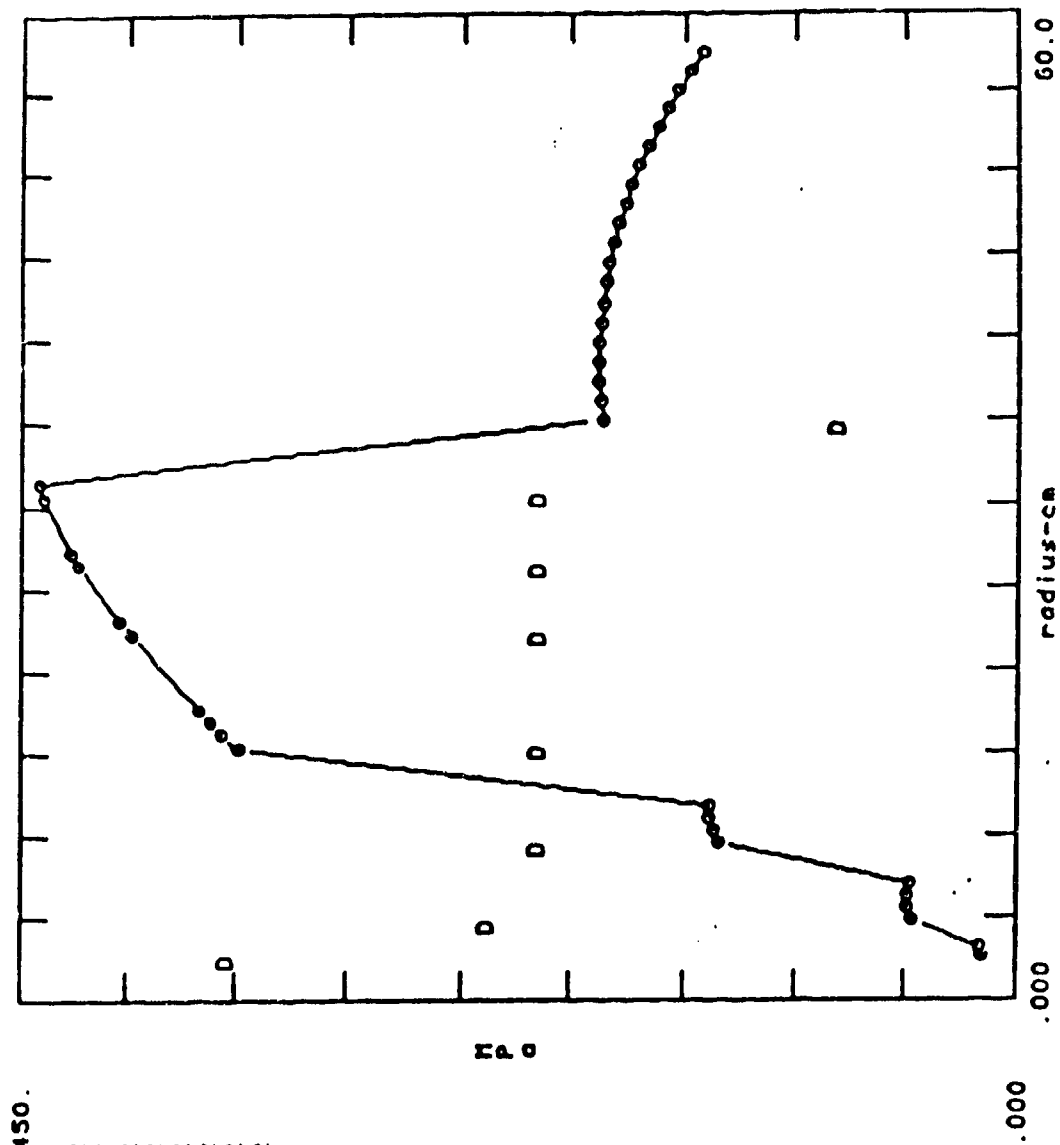
Percentage Radial Displacement from Hoop Stress

Elongation

bore = 19.3416
 wt. cond = 5.95717 ton
 wt. sleeve = .497680 ton
 I layers = 40
 I turns = 1306 x 2
 layer 1 = 2
 layer 2 = 4
 layer 3 = 4
 layer 4 = 4
 layer 5 = 2
 layer 6 = 2
 layer 7 = 2
 layer 8 = 20
 turns 1 = 12 x 2
 turns 2 = 18 x 2
 turns 3 = 30 x 2
 turns 4 = 45 x 2
 turns 5 = 45 x 2
 turns 6 = 45 x 2
 turns 7 = 45 x 2
 turns 8 = 32 x 2



Total Axial Stress at Midplane (w/o load sharing)



Axial Stress 450.

- J 1 = 51.1375 Kamps/cm²
- J 2 = 34.4587 Kamps/cm²
- J 3 = 31.0841 Kamps/cm²
- J 4 = 31.0841 Kamps/cm²
- J 5 = 31.0841 Kamps/cm²
- J 6 = 31.0841 Kamps/cm²
- J 7 = 31.0841 Kamps/cm²
- J 8 = 11.7388 Kamps/cm²

D = relative current density

Fig. 8

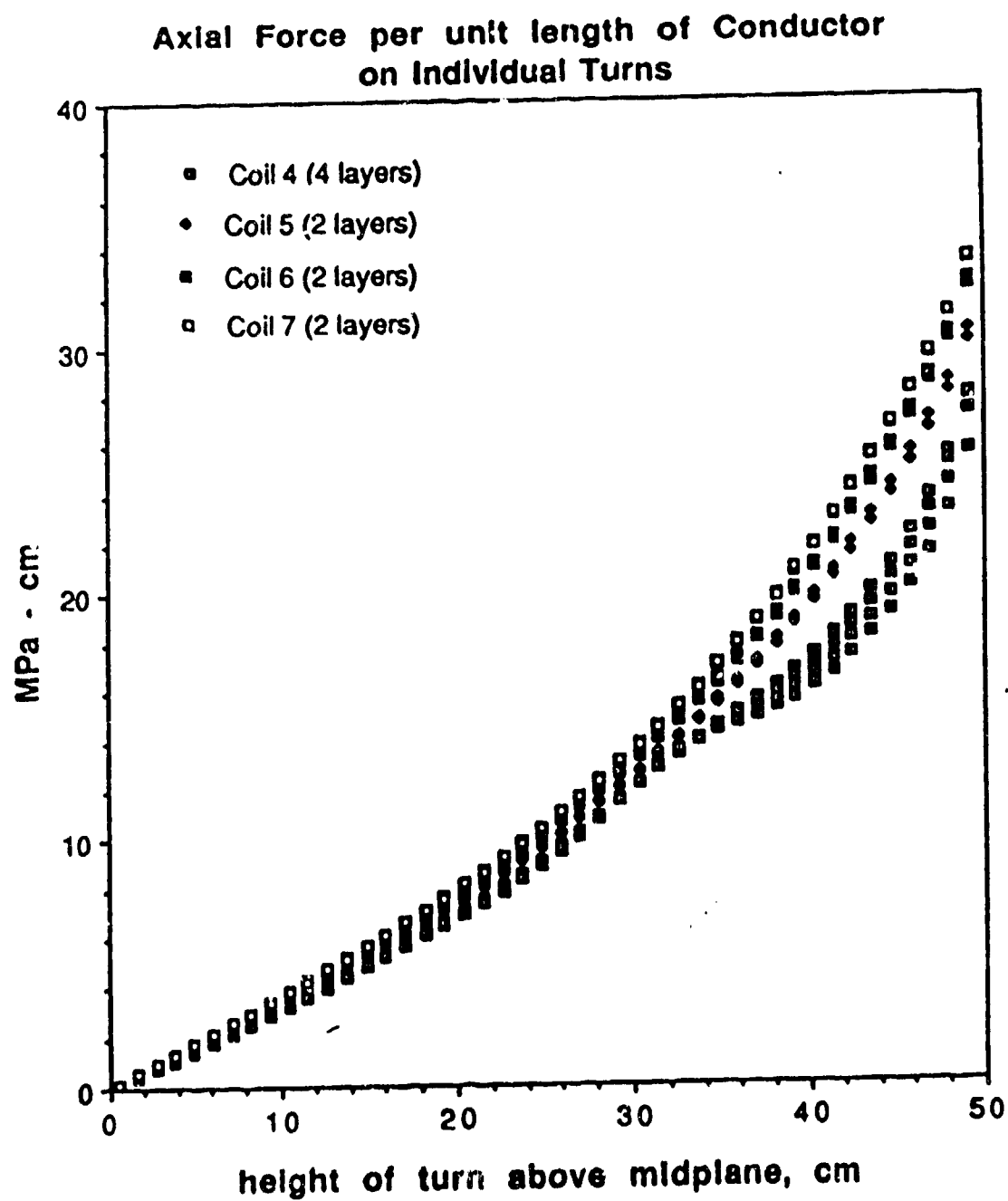


Fig. 9

INTERFACE PRESSURE AND THERMAL STRESS

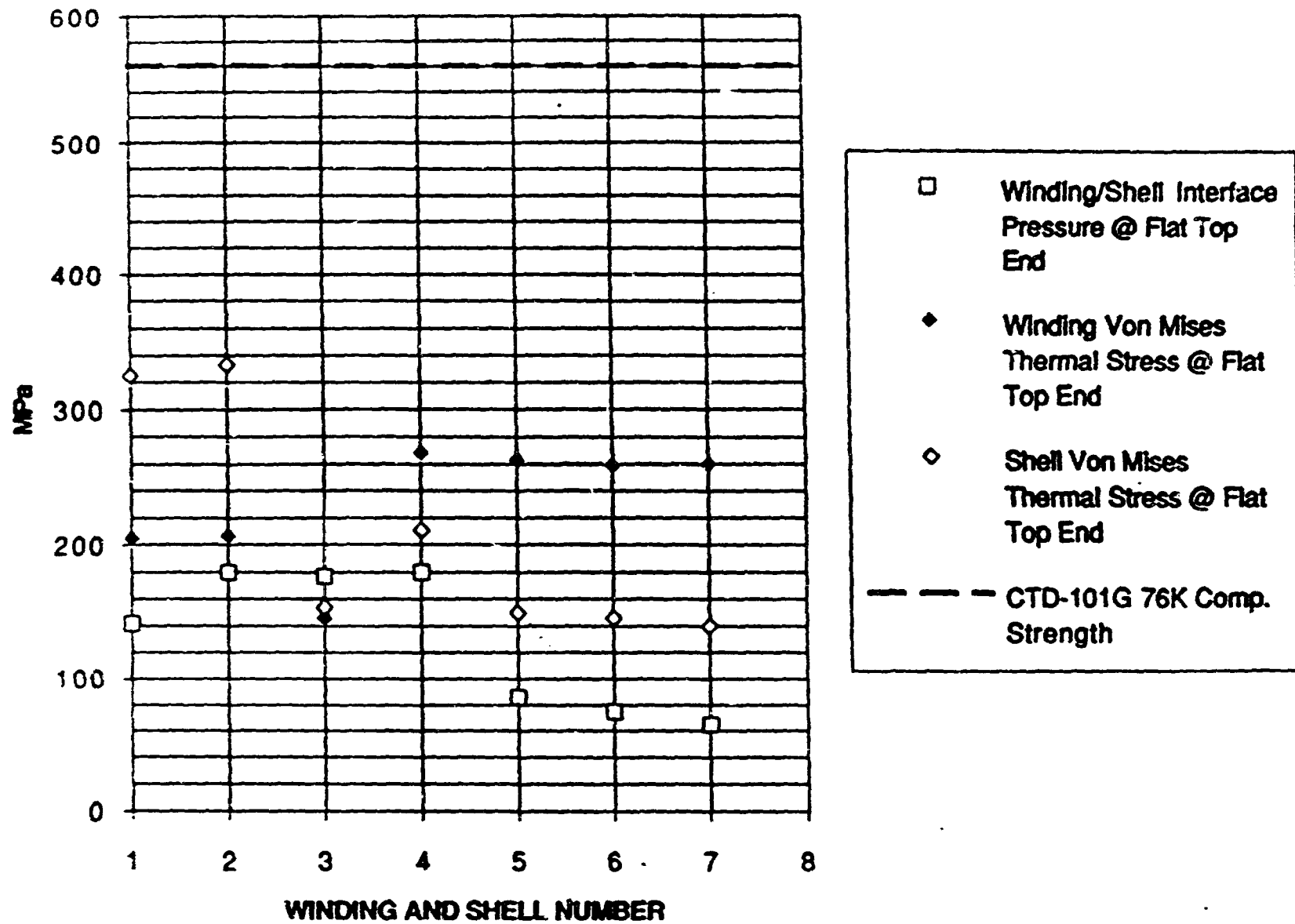
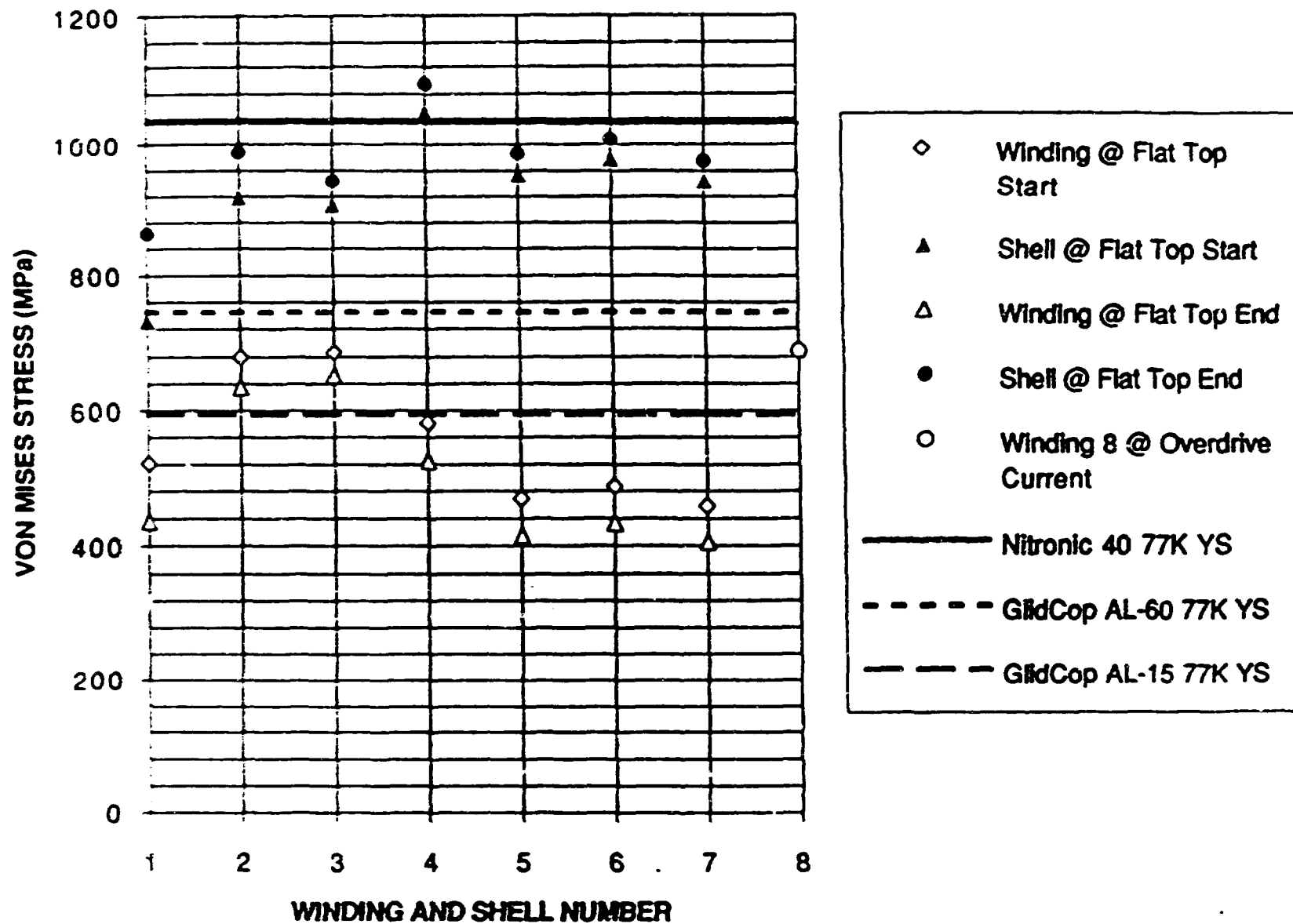


Fig. 10

WINDING AND SHELL VON MISES STRESS (on horiz. mid-plane at Inner diameter)



Coil Lengths Compared to Saturation and Homogeneity Lengths

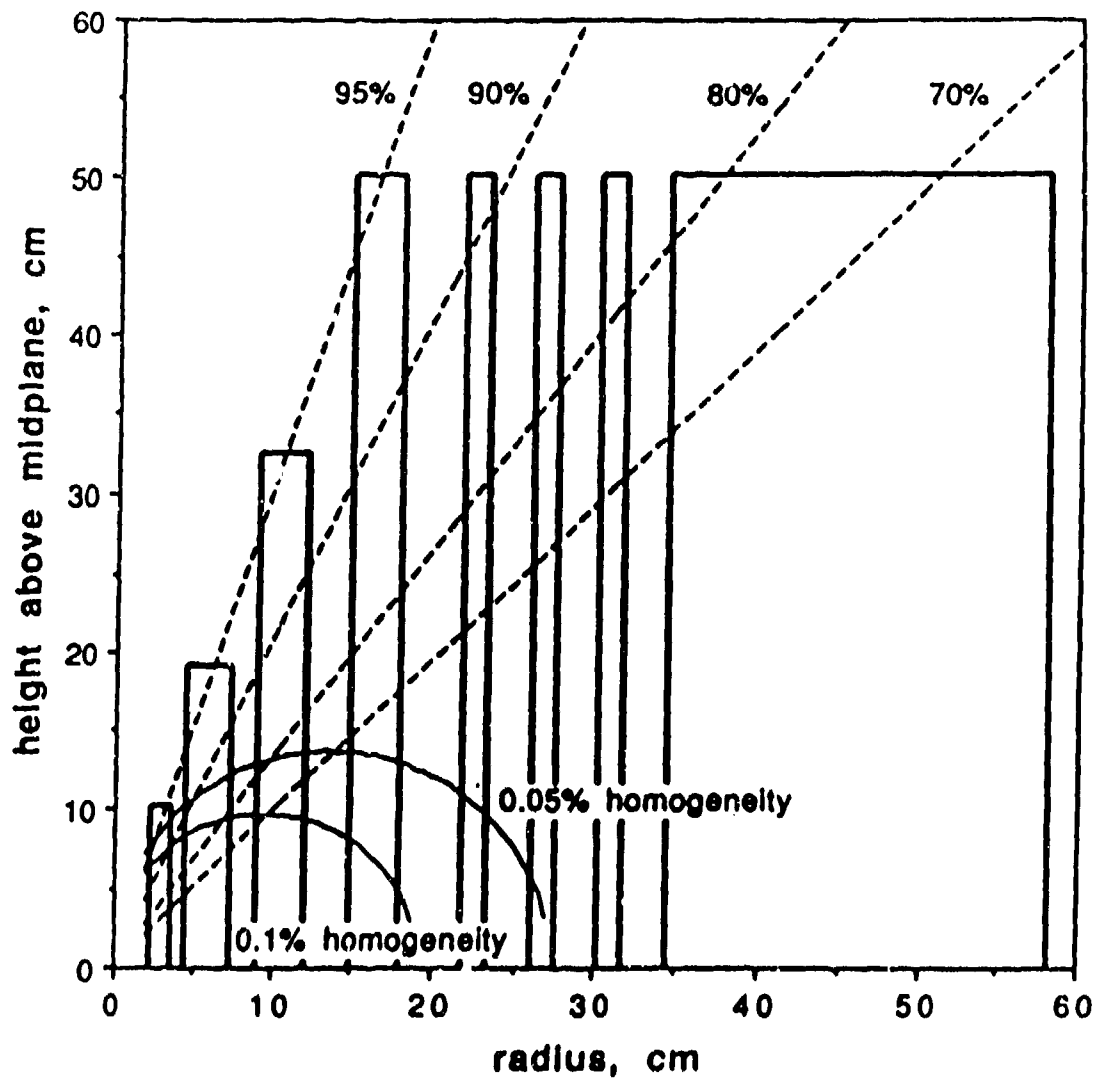
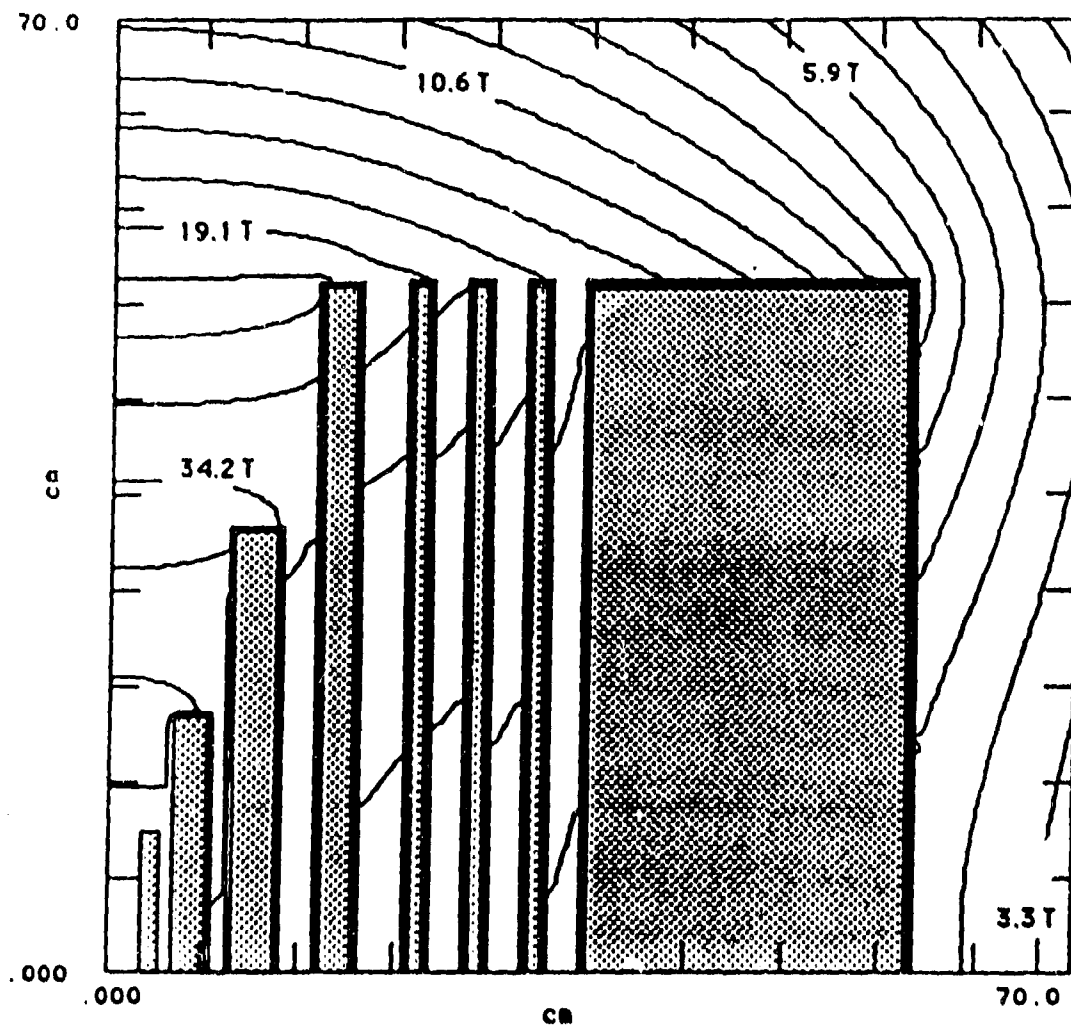


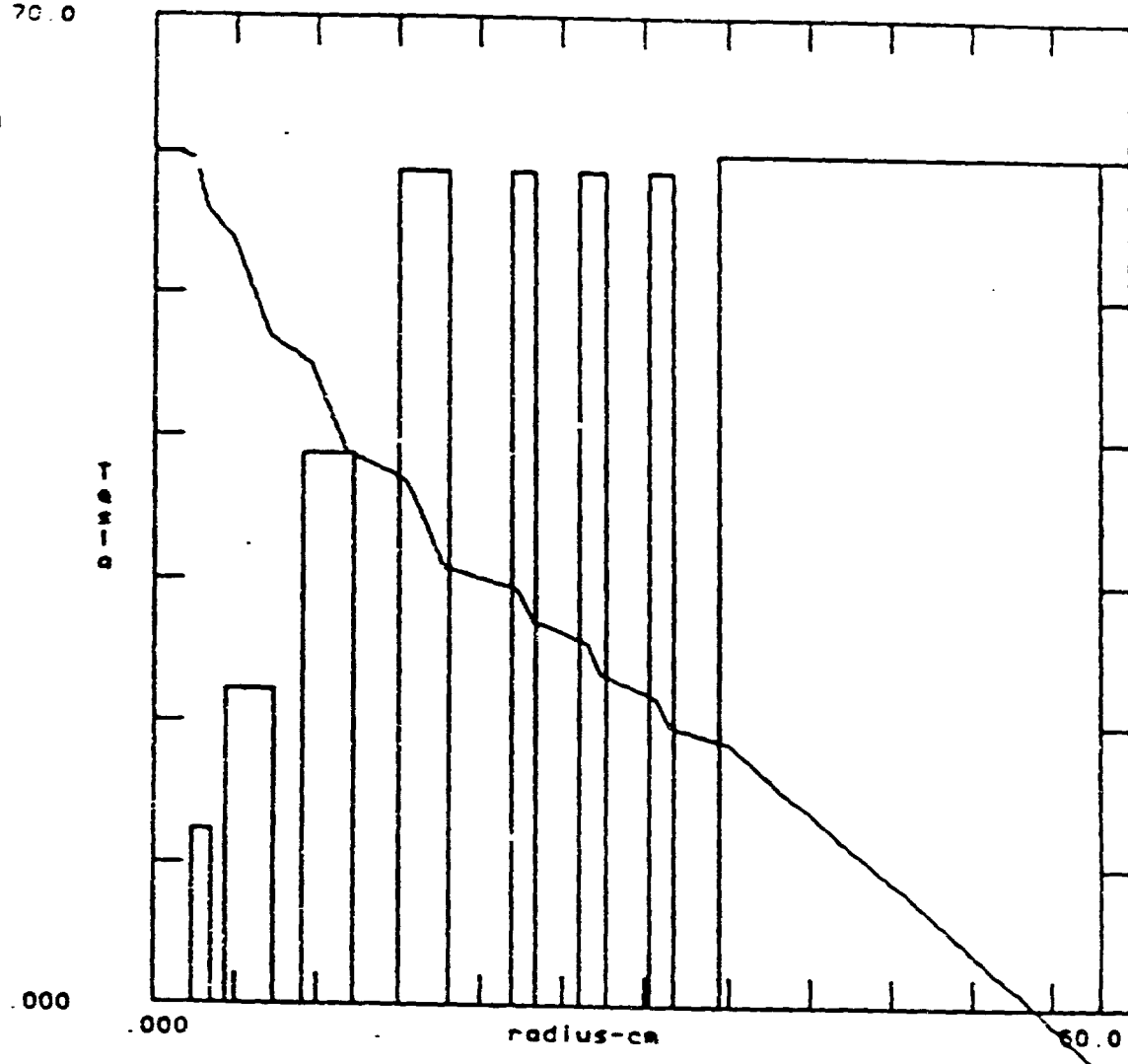
Fig. 1.

Contours of Constant Field

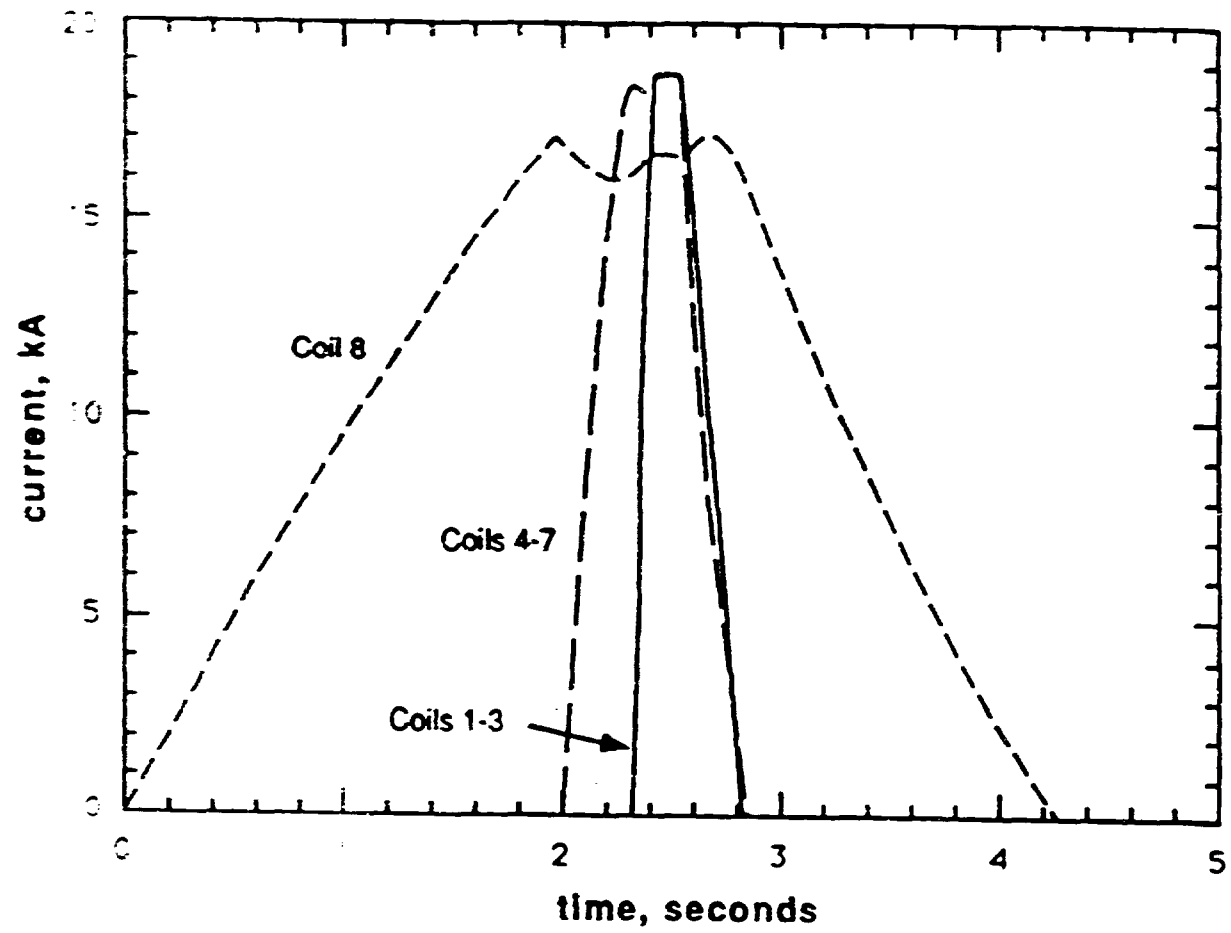


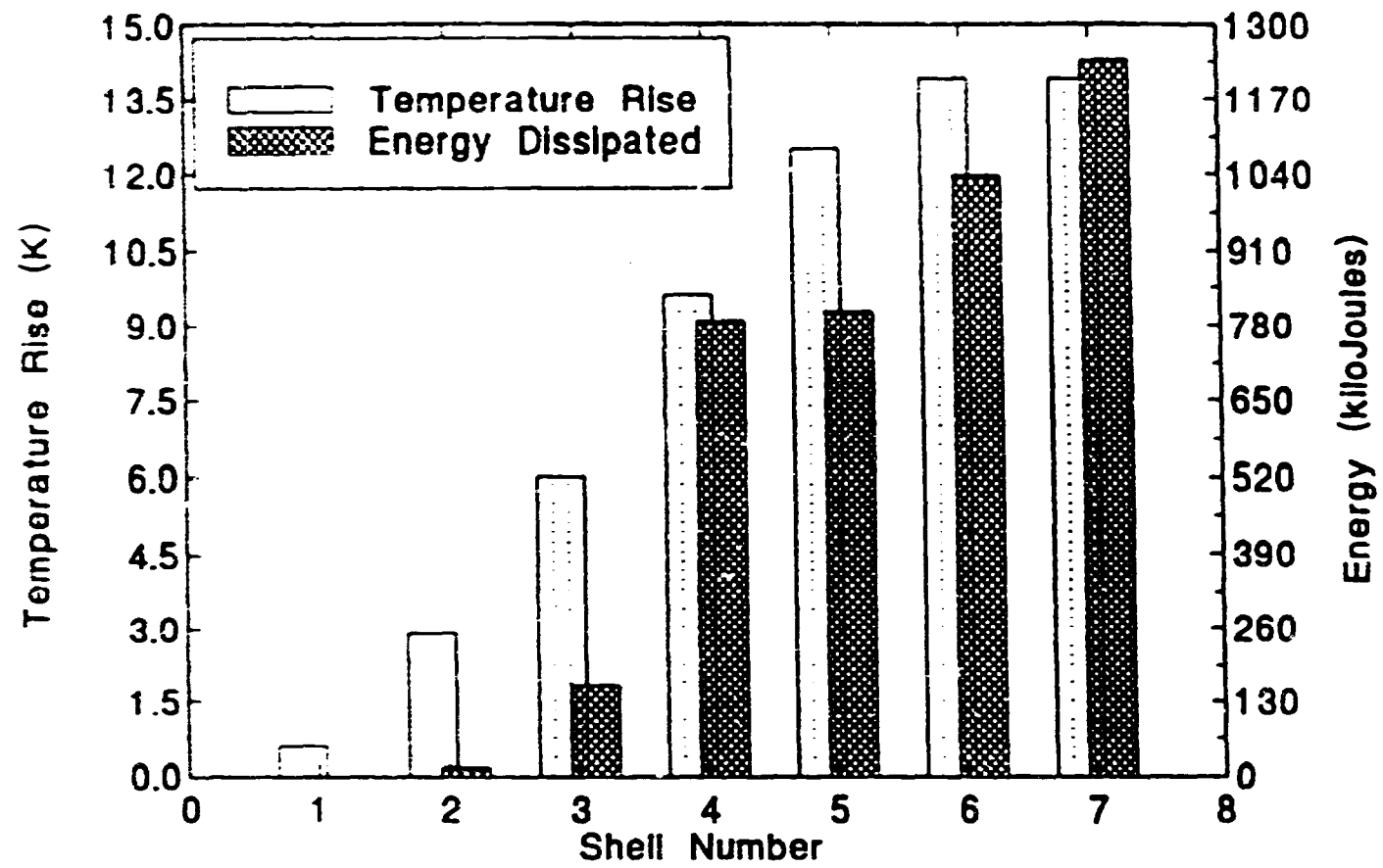
Magnetic Field Profile

B_z 70.0
 R -cm = 1.900000E+06
 ρ cm = 301.708
 $B_z(0)$ = 60.0000 tesla
 B homog = 1.222985E-04
 r sphr = 1.00000 cm
 B_z 1 = 5.13833
 B_z 2 = 8.45280
 B_z 3 = 8.06179
 B_z 4 = 7.99207
 B_z 5 = 3.83192
 B_z 6 = 3.70575
 B_z 7 = 3.57298
 B_z 8 = 19.2244
 power 1 = 1.613963E+06
 power 2 = 6.547257E+06
 power 3 = 1.754704E+07
 power 4 = 4.074007E+07
 power 5 = 2.793781E+07
 power 6 = 3.308587E+07
 power 7 = 3.823394E+07
 power 8 = 1.350017E+08

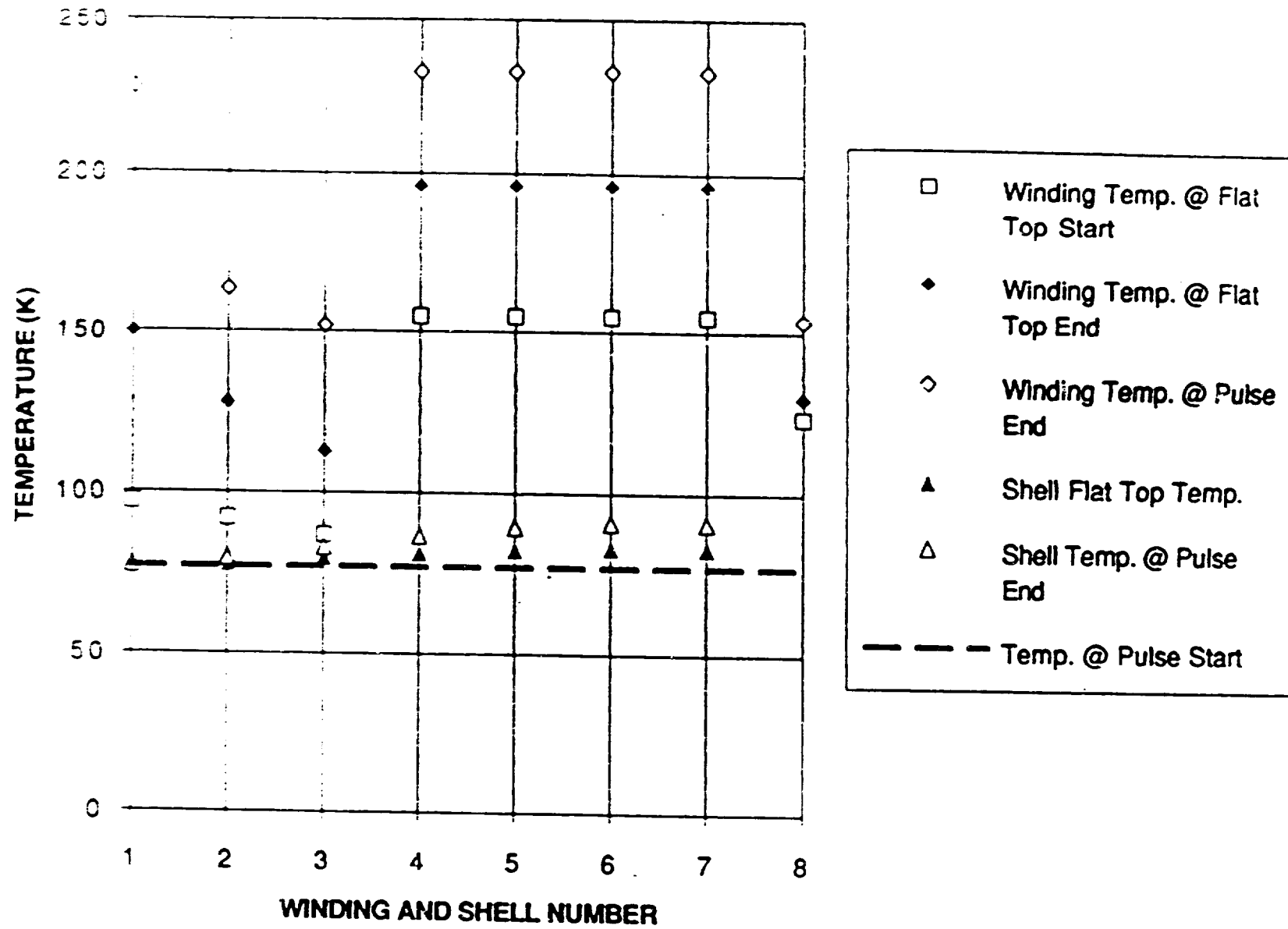


Current Profile for 100 ms Flat-top Pulse

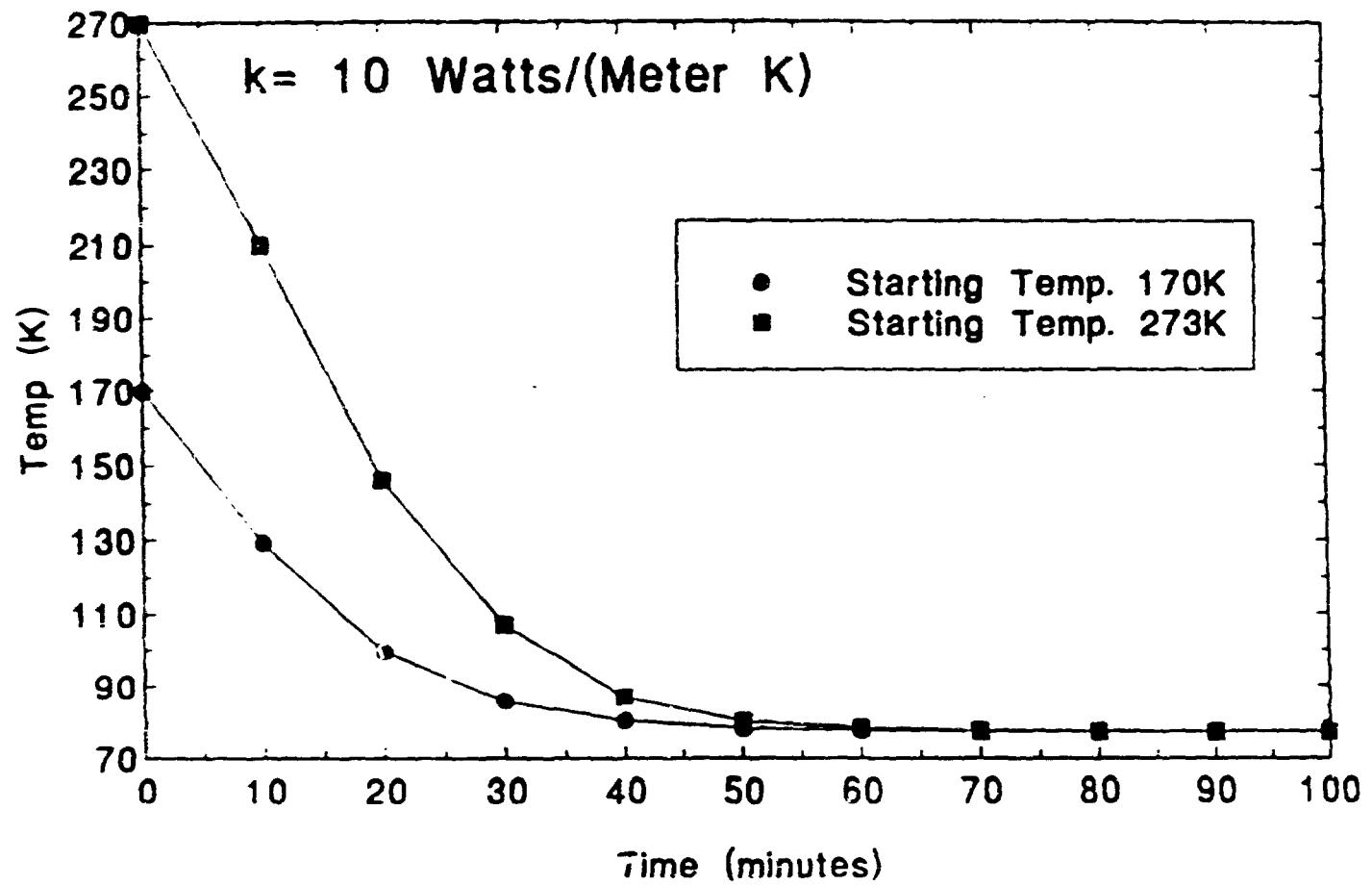




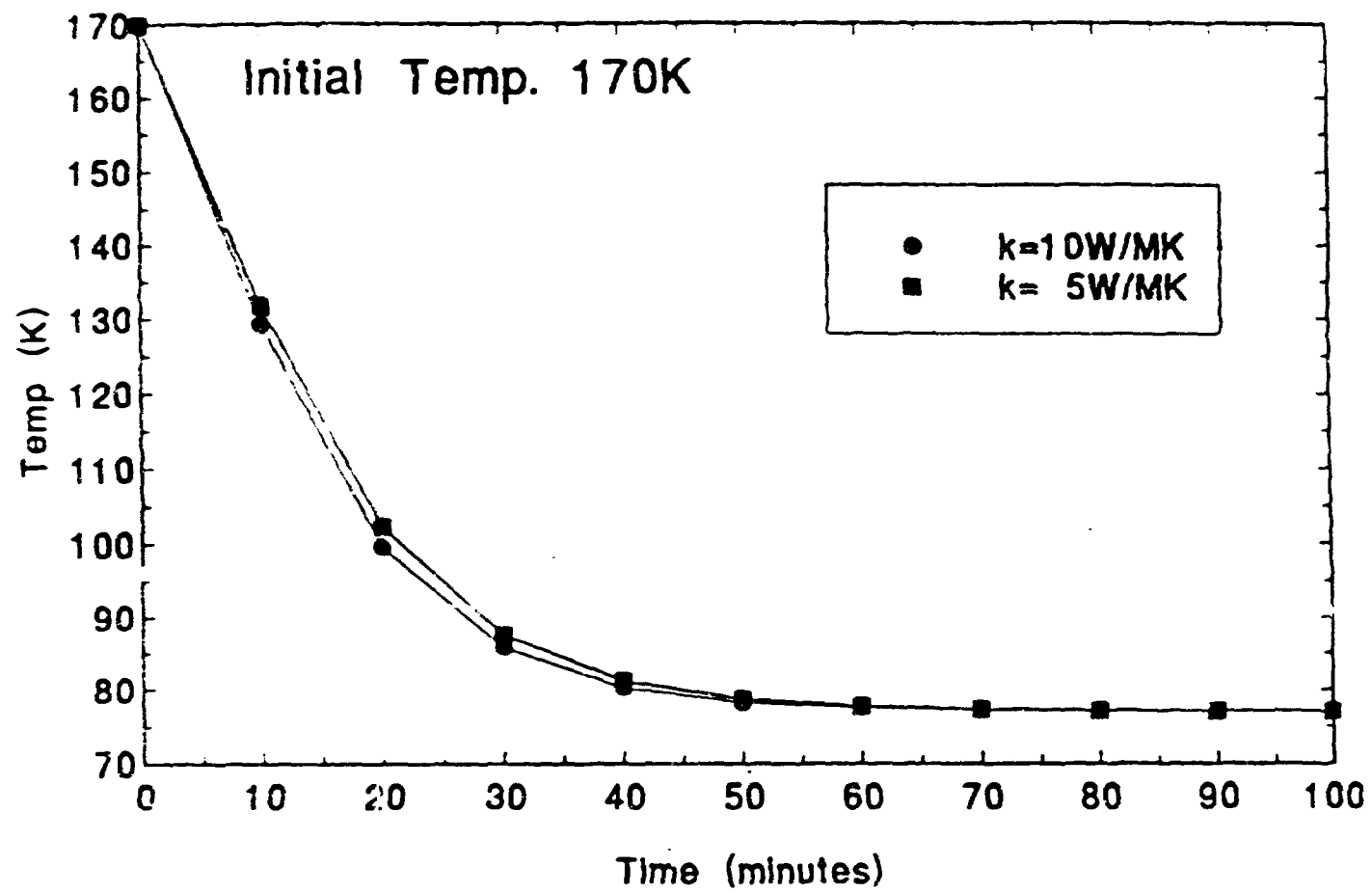
WINDING AND SHELL TEMPERATURE



OUTER COIL COOL-DOWN



OUTER COIL COOL-DOWN



COIL COOL-DOWN

